

ON Semiconductor®

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Bi-CMOS LSI

Fan Motor Driver for PC and Server

Overview

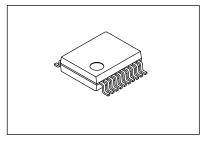
The LV8805SV is a motor driver for PC and server fans.

Feature

• Direct PWM three-phase sensorless motor driver.

Typical Applications

- PC and Server
- Refrigerator
- Server
- Desktop Computer



SSOP20J (225mil)

Specifications

Absolute Maximum Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
V _{CC} maximum supply voltage	V _{CC} max		16	V
VG maximum supply voltage	VG max		21	V
OUT pin withstand voltage	V _{OUT} max		16	V
OUT pin maximum output	I _{OUT} max	UO pin, VO pin, WO pin	1.2	Α
SOFTST pin withstand voltage	V _{SOFTST} max		6	V
FR pin withstand voltage	V _{FR} max		6	V
PWMIN pin withstand voltage	V _{PWMIN} max		6	V
FG output pin withstand voltage	V _{FG} max		16	V
FG pin output current	I _{FG} max		5	mA
RD output pin withstand voltage	VRD max		16	V
RD pin output current	IRD max		5	mA
Allowable Power dissipation 1	Pd max1	Independent IC	0.3	W
Allowable Power dissipation 2	Pd max2	Mounted on designated board *1	0.95	W
Operating temperature	Topr		-40 to +95	°C
Storage temperature	Tstg	*2	–55 to +150	°C

^{*1:} When mounted on the designated 76.1mm × 114.3mm × 1.6mm, glass epoxy board (single-layer)

Caution 1) Absolute maximum ratings represent the value which cannot be exceeded for any length of time.

Caution 2) Even when the device is used within the range of absolute maximum ratings, as a result of continuous usage under high temperature, high current, high voltage, or drastic temperature change, the reliability of the IC may be degraded. Please contact us for the further details.

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

ORDERING INFORMATION

See detailed ordering and shipping information on page 11 of this data sheet.

^{*2:} Do not exceed Tjmax=150°C.

Recommended Operating Conditions at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
V _{CC} supply voltage	V _{CC}		6 to 15	V
SOFTST input voltage range	VSOFTST		0 to VREG	V
FR input voltage range	V_{FR}		0 to VREG	V
MINSP input voltage range	V _{MINSP}		0 to VREG	V

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

Electrical Characteristics at Ta = 25°C, V_{CC} = 12V, unless otherwise specified

Parameter	Symbol	Symbol Conditions	Ratings			Unit
Falanielei	, , , , , , , , , , , , , , , , , , ,	Conditions	min	typ	max	Office
Circuit current 1	I _{CC} 1			2.6	3.6	mA
Charge pump block						
Charge pump output voltage	V _{VG}			17		V
Regulator block						
5V regulator voltage	V _{VREG}		4.75	5	5.25	V
Output on resistance						
Sum of high-/low-side output transistor on resistance	Ron (H+L)	I _O = 0.7A, V _{CC} = 12V, VG = 17V		1.2	2	Ω
Startup oscillator (OSC) pin						
OSC pin charge current	I _{OSC} C			-2.5		μА
OSC pin discharge current	I _{OSC} D			2.5		μΑ
PWM input (PWMIN) pin	- 1	•				
High-level input voltage range	V _{PWMIN} H		2.3		VREG	V
Low-level input voltage range	V _{PWMIN} L		0		1	V
Range of PWM input frequency	fPWMIN		15		60	kHz
Minimum pulse width	T _{MINPW}	Input HIGH voltage 5[V] and input LOW voltage 0[V]			0.2	μS
Duty cycle range is determined by $(T_{MINPW} \times f_{PWMIN}) \times 100\% \text{ for minimu} \\ (1 - T_{MINPW} \times f_{PWMIN}) \times 100\% \text{ for max} \\ \text{When } f_{PWMIN} = 60[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], \text{ the input PWM duty c} \\ \text{When } f_{PWMIN} = 50[\text{kHz}], the input PWM dut$	kimum ycle range = 1.2%					
$(T_{MINPW} \times f_{PWMIN}) \times 100\%$ for minimu $(1 - T_{MINPW} \times f_{PWMIN}) \times 100\%$ for max When $f_{PWMIN} = 60$ [kHz], the input PWM duty c	kimum ycle range = 1.2% ycle range = 1.0% ycle range = 0.5%	- 99.0% - 99.5%				
$(T_{MINPW} \times f_{PWMIN}) \times 100\%$ for minimu $(1 - T_{MINPW} \times f_{PWMIN}) \times 100\%$ for max When $f_{PWMIN} = 60$ [kHz], the input PWM duty c When $f_{PWMIN} = 50$ [kHz], the input PWM duty c When $f_{PWMIN} = 25$ [kHz], the input PWM duty c	kimum ycle range = 1.2% ycle range = 1.0% ycle range = 0.5%	- 99.0% - 99.5%				
$(T_{MINPW} \times f_{PWMIN}) \times 100\%$ for minimu $(1 - T_{MINPW} \times f_{PWMIN}) \times 100\%$ for max When $f_{PWMIN} = 60$ [kHz], the input PWM duty c When $f_{PWMIN} = 50$ [kHz], the input PWM duty c When $f_{PWMIN} = 25$ [kHz], the input PWM duty c When $f_{PWMIN} = 15$ [kHz], the input PWM duty c	kimum ycle range = 1.2% ycle range = 1.0% ycle range = 0.5%	- 99.0% - 99.5%	2.3		VREG	V
$(T_{MINPW} \times f_{PWMIN}) \times 100\%$ for minimu $(1 - T_{MINPW} \times f_{PWMIN}) \times 100\%$ for max When $f_{PWMIN} = 60[\text{kHz}]$, the input PWM duty of When $f_{PWMIN} = 50[\text{kHz}]$, the input PWM duty of When $f_{PWMIN} = 25[\text{kHz}]$, the input PWM duty of When $f_{PWMIN} = 15[\text{kHz}]$, the input PWM duty of Forward/reverse switching pin	dimum ycle range = 1.2% ycle range = 1.0% ycle range = 0.5% ycle range = 0.3%	- 99.0% - 99.5% - 99.7% Order of current application :	2.3		VREG	V
$(T_{MINPW} \times f_{PWMIN}) \times 100\%$ for minimu $(1 - T_{MINPW} \times f_{PWMIN}) \times 100\%$ for maximum when $f_{PWMIN} = 60$ [kHz], the input PWM duty c When $f_{PWMIN} = 50$ [kHz], the input PWM duty c When $f_{PWMIN} = 25$ [kHz], the input PWM duty c When $f_{PWMIN} = 15$ [kHz], the input PWM duty c Forward/reverse switching pin High-level input voltage range	kimum ycle range = 1.2% ycle range = 1.0% ycle range = 0.5% ycle range = 0.3% VFRH	- 99.0% - 99.5% - 99.7% Order of current application: UOUT→VOUT→WOUT Order of current application:				
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(T _{MINPW} x f _{PWMIN}) x 100% for minimu (1 - T _{MINPW} x f _{PWMIN}) x 100% for max When f _{PWMIN} = 60[kHz], the input PWM duty c When f _{PWMIN} = 50[kHz], the input PWM duty c When f _{PWMIN} = 25[kHz], the input PWM duty c When f _{PWMIN} = 15[kHz], the input PWM duty c Forward/reverse switching pin High-level input voltage range Low-level input voltage range FG and RD output pins FG output pin low-level voltage FG output pin leak voltage	wimum yycle range = 1.2% yycle range = 1.0% yycle range = 0.5% yycle range = 0.3% VFRH VFRL VFG ILFG	- 99.0% - 99.5% - 99.7% Order of current application: UOUT→VOUT→WOUT Order of current application: UOUT→WOUT→VOUT When I _Q is 2mA When V _{FG} is 16V			0.35	V V μΑ
(T _{MINPW} x f _{PWMIN}) x 100% for minimu (1 - T _{MINPW} x f _{PWMIN}) x 100% for max When f _{PWMIN} = 60[kHz], the input PWM duty c When f _{PWMIN} = 50[kHz], the input PWM duty c When f _{PWMIN} = 25[kHz], the input PWM duty c When f _{PWMIN} = 15[kHz], the input PWM duty c Forward/reverse switching pin High-level input voltage range Low-level input voltage range FG and RD output pins FG output pin low-level voltage RD output pin low-level voltage	vigle range = 1.2% ycle range = 1.0% ycle range = 0.5% ycle range = 0.3% VFRH VFRL VFG ILFG VRD	- 99.0% - 99.5% - 99.7% Order of current application: UOUT→VOUT→WOUT Order of current application: UOUT→WOUT→VOUT When I _O is 2mA When V _{FG} is 16V When I _O is 2mA			0.35 1 0.35	V V µА
(T _{MINPW} x f _{PWMIN}) x 100% for minimu (1 - T _{MINPW} x f _{PWMIN}) x 100% for max When f _{PWMIN} = 60[kHz], the input PWM duty c When f _{PWMIN} = 55[kHz], the input PWM duty c When f _{PWMIN} = 25[kHz], the input PWM duty c When f _{PWMIN} = 15[kHz], the input PWM duty c Forward/reverse switching pin High-level input voltage range Low-level input voltage range FG and RD output pins FG output pin low-level voltage FG output pin leak voltage RD output pin leak voltage	vigle range = 1.2% ycle range = 1.0% ycle range = 0.5% ycle range = 0.3% VFRH VFRL VFG ILFG VRD	- 99.0% - 99.5% - 99.7% Order of current application: UOUT→VOUT→WOUT Order of current application: UOUT→WOUT→VOUT When I _O is 2mA When V _{FG} is 16V When I _O is 2mA			0.35 1 0.35	V V LIA
(T _{MINPW} x f _{PWMIN}) x 100% for minimu (1 - T _{MINPW} x f _{PWMIN}) x 100% for max When f _{PWMIN} = 60[kHz], the input PWM duty c When f _{PWMIN} = 25[kHz], the input PWM duty c When f _{PWMIN} = 15[kHz], the input PWM duty c When f _{PWMIN} = 15[kHz], the input PWM duty c Forward/reverse switching pin High-level input voltage range Low-level input voltage range FG and RD output pins FG output pin low-level voltage FG output pin leak voltage RD output pin leak voltage RD output pin leak voltage Current limiter circuit	vertical line vertical lin	- 99.0% - 99.5% - 99.7% Order of current application: UOUT→VOUT→WOUT Order of current application: UOUT→WOUT→VOUT When I _O is 2mA When V _{FG} is 16V When I _O is 2mA When V _{RD} is 16V	0	0.25	0.35 1 0.35 1	V V µA V
(T _{MINPW} x f _{PWMIN}) x 100% for minimu (1 - T _{MINPW} x f _{PWMIN}) x 100% for max When f _{PWMIN} = 60[kHz], the input PWM duty c When f _{PWMIN} = 55[kHz], the input PWM duty c When f _{PWMIN} = 25[kHz], the input PWM duty c When f _{PWMIN} = 15[kHz], the input PWM duty c Forward/reverse switching pin High-level input voltage range Low-level input voltage range FG and RD output pins FG output pin low-level voltage FG output pin low-level voltage RD output pin leak voltage RD output pin leak voltage Current limiter circuit Limiter voltage	vertical line vertical lin	- 99.0% - 99.5% - 99.7% Order of current application: UOUT→VOUT→WOUT Order of current application: UOUT→WOUT→VOUT When I _O is 2mA When V _{FG} is 16V When I _O is 2mA When V _{RD} is 16V	0	0.25	0.35 1 0.35 1	V V µA V
(T _{MINPW} x f _{PWMMN}) x 100% for minimu (1 - T _{MINPW} x f _{PWMMN}) x 100% for max When f _{PWMMN} = 60[kHz], the input PWM duty c When f _{PWMMN} = 50[kHz], the input PWM duty c When f _{PWMMN} = 25[kHz], the input PWM duty c When f _{PWMMN} = 15[kHz], the input PWM duty c Forward/reverse switching pin High-level input voltage range Low-level input voltage range FG and RD output pins FG output pin low-level voltage FG output pin leak voltage RD output pin leak voltage RD output pin leak voltage Current limiter circuit Limiter voltage Constraint protection circuit	VFR	- 99.0% - 99.5% - 99.7% Order of current application: UOUT→VOUT→WOUT Order of current application: UOUT→WOUT→VOUT When I _O is 2mA When V _{FG} is 16V When I _O is 2mA When V _{RD} is 16V	0.225	0.25	0.35 1 0.35 1 0.275	V µА V µА
(T _{MINPW} x f _{PWMMN}) x 100% for minimu (1 - T _{MINPW} x f _{PWMMN}) x 100% for max When f _{PWMMN} = 60[kHz], the input PWM duty c When f _{PWMMN} = 55[kHz], the input PWM duty c When f _{PWMMN} = 25[kHz], the input PWM duty c When f _{PWMMN} = 15[kHz], the input PWM duty c Forward/reverse switching pin High-level input voltage range Low-level input voltage range FG and RD output pins FG output pin low-level voltage FG output pin leak voltage RD output pin leak voltage RD output pin leak voltage Current limiter circuit Limiter voltage Constraint protection circuit CT pin high-level voltage	winum ycle range = 1.2% ycle range = 1.0% ycle range = 0.5% ycle range = 0.3% VFRH VFRL VFG ILFG VRD ILRD VRF	- 99.0% - 99.5% - 99.7% Order of current application: UOUT→VOUT→WOUT Order of current application: UOUT→WOUT→VOUT When I _O is 2mA When V _{FG} is 16V When I _O is 2mA When V _{RD} is 16V	0.225	0.25	0.35 1 0.35 1 0.275	V μΑ V μΑ
(T _{MINPW} x f _{PWMMN}) x 100% for minimu (1 - T _{MINPW} x f _{PWMMN}) x 100% for max When f _{PWMMN} = 60[kHz], the input PWM duty c When f _{PWMMN} = 55[kHz], the input PWM duty c When f _{PWMMN} = 15[kHz], the input PWM duty c When f _{PWMMN} = 15[kHz], the input PWM duty c Forward/reverse switching pin High-level input voltage range Low-level input voltage range FG and RD output pins FG output pin low-level voltage FG output pin leak voltage RD output pin leak voltage RD output pin leak voltage Current limiter circuit Limiter voltage Constraint protection circuit CT pin high-level voltage CT pin low-level voltage	VFR	- 99.0% - 99.5% - 99.7% Order of current application: UOUT→VOUT→WOUT Order of current application: UOUT→WOUT→VOUT When I _O is 2mA When V _{FG} is 16V When I _O is 2mA When V _{RD} is 16V	0.225 2.25 0.43	0.25 0.25 2.8 0.5	0.35 1 0.35 1 0.275	V

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Description	0	Conditions	Ratings			11.2
Parameter	Symbol		min	typ	max	Unit
Soft start circuit						
Soft start releasing voltage	VSOFTST			2.5		V
SOFTST pin charge current	ISOFTST			0.6		μА
Thermal protection circuit						
Thermal protection circuit operating TSD		Design target *	150	180	210	°C
temperature						ĺ

^{*:} Design target value and no measurement is made. The thermal protection circuit is incorporated to protect the IC from burnout or thermal destruction. Since it operates outside the IC's guaranteed operating range, the customer's thermal design should be performed so that the thermal protection circuit will not be activated when the fan is running under normal operating conditions.

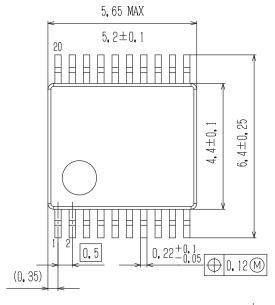
Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

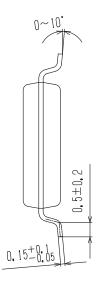
Package Dimensions

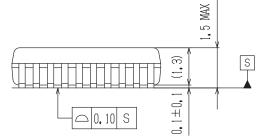
unit: mm

SSOP20J (225mil)

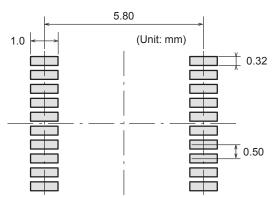
CASE 565AP ISSUE A







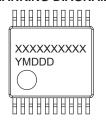
SOLDERING FOOTPRINT*



NOTE: The measurements are not to guarantee but for reference only.

*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

GENERIC MARKING DIAGRAM*



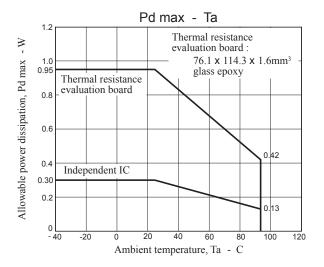
XXXXX = Specific Device Code

Y = Year

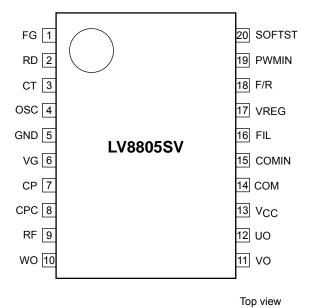
M = Month

DDD = Additional Traceability Data

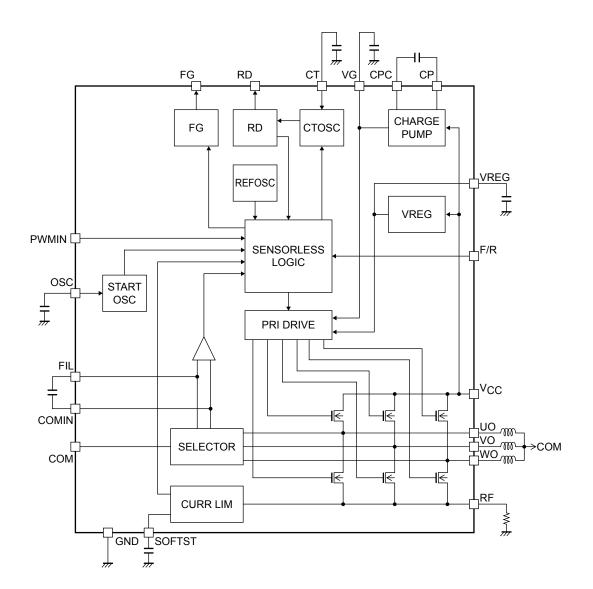
^{*}This information is generic. Please refer to device data sheet for actual part marking. Pb–Free indicator, "G" or microdot " ■", may or may not be present.



Pin Assignment



Block Diagram



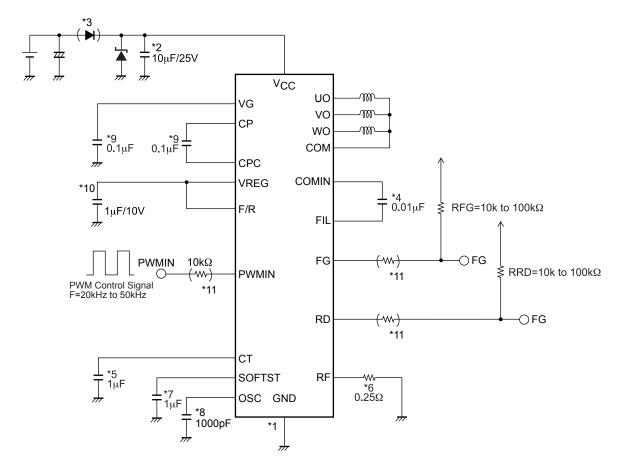
Pin Function

FG	FG pulse output. This pin outputs a Hall	
	sensor system equivalent pulse signal.	12
RD	Motor lockup detection output. Output is fixed high when motor is locked up.	
СТ	Motor lockup detection time setting. When the motor lockup condition is detected, the protection time period before the protection circuit is activated is set by connecting a capacitor between this pin and ground.	VREG
OSC	Motor startup frequency setting. A capacitor must be connected between this pin and ground. The startup frequency is adjusted by controlling the charge/discharge current and capacitance of the capacitor.	VREG 50000 W 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7
GND	GND pin	
VG	Charge pump step-up voltage output. A capacitor must be connected between this pin and the V _{CC} pin or ground.	7 \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \
СР	Charge pump step-up pulse output pin. A capacitor must be connected between this pin and the CPC pin (pin 8).	VREG 6
CPC	Charge pump step-up pin. A capacitor must be connected between this pin and the CP pin (pin 7).	
V _{CC}	Power supply for the IC and motor. Capacitors must be connected between these pins and ground.	
UO	Output pins. Connect these pins to the U, V,	
VO	and W of the motor coil.	
WO RF	Output current detection pins. The drive current is detected by connecting a resistor between these pins and ground.	9
	OSC GND VG CP CPC UO VO WO	CT Motor lockup detection time setting. When the motor lockup condition is detected, the protection time period before the protection circuit is activated is set by connecting a capacitor between this pin and ground. Motor startup frequency setting. A capacitor must be connected between this pin and ground. The startup frequency is adjusted by controlling the charge/discharge current and capacitance of the capacitor. GND GND pin VG Charge pump step-up voltage output. A capacitor must be connected between this pin and the V _{CC} pin or ground. CP Charge pump step-up pulse output pin. A capacitor must be connected between this pin and the CPC pin (pin 8). CPC Charge pump step-up pin. A capacitor must be connected between this pin and the CPC pin (pin 7). VCC Power supply for the IC and motor. Capacitors must be connected between this pin and the CP pin (pin 7). VCC Power supply for the IC and motor. Capacitors must be connected between these pins and ground. UO Output pins. Connect these pins to the U, V, and W of the motor coil. VC RF Output current detection pins. The drive current is detected by connecting a resistor

Continued on next page.

Pin No.	from preceding Pin name	Function	Equivalent circuit
			·
14 15	COMIN	Motor middle point connection. Motor position detection comparator filter pin. A capacitor must be connected between this pin and the FIL pin (pin 16).	VG TWI THE TWO
16	FIL	Motor position detection comparator filter pin. A capacitor must be connected between this pin and the COMIN pin (pin 15).	15 16
17	VREG	Regulator voltage (5V) output. A capacitor must be connected between these pins and ground.	VCC TO
18	F/R	Motor rotation direction switching. A high-level input causes current to flow into the motor in the order of U, V, and W and a low-level input in the order of U, W, and V. Changing the order of current application turns the motor in the opposite direction.	VREG Reverse signal 15kΩ Forward/reverse switching signal Forward signal
19	PWMIN	PWM signal input pin. "H" The output transistor is turned on by the level voltage input. "L" The output transistor is turned off by the level voltage input, and the motor stops. The speed of the motor is controlled by controlling Duty of the input signal. When the pin opens, the motor becomes all velocities.	VREG \$300kΩ 5 15 15 15 15 15 15 15
20	SOFTST	Soft start time setting. The motor can be started smoothly by connecting a capacitor between this pin and ground.	VREG 500Ω\$ 20 11 11 11 11 11 11 11 11 11

Application Circuit Example



*1. Power supply and GND wiring

The GND is connected to the control circuit power supply system.

*2. Power-side power stabilization capacitor

For the power-side power stabilization capacitor, use a capacitor of 10µF or more.

Connect the capacitor between V_{CC} and GND with a thick and along the shortest possible route.

LV8805SV uses synchronous rectification for high efficiency drive. Synchronous rectification is effective for heat reduction and higher efficiency. However, it may increase supply voltage.

If the supply voltage shall increase, make sure that it does not exceed the maximum ratings by inserting a zener diode between power supply and GND.

*3. Reverse connection protection diode

This diode protects reverse connection.

Insert a diode between power supply and V_{CC} pin to protect the IC from destruction due to reverse connection. Connection of this diode is not necessary required.

*4. COMIN and FIL pins

These pins are used to connect the filter capacitor. The LV8805SV uses the back EMF signal generated when the motor is running to detect the information on the rotor position. The IC determines the timing at which the output block applies current to the motor based on the position information obtained here. Insert a filter capacitor with a capacitance ranging from (1,000pF to 10,000pF) between the COMIN pin and FIL pin to prevent any motor startup misoperation that is caused by noise. However, care must be taken since an excessively high capacitance will give rise to deterioration in efficiency and delays in the output power-on timing while the motor is running at high speed. Furthermore, connect the capacitor between the COMIN pin and FIL pin as close as possible in order to avoid the effects of noise from other sources.

*5. CT pin

This pin is used to connect the lock detection capacitor.

The constant-current charging and constant-current discharging circuits incorporated cause locking when the pin voltage reaches 2.5V, and releasing the lock protection when it drops to 0.5V. This pin must be connected to the GND when it is not going to be used.

*6. RF pins

These pins are used to set the current limit.

When the pin voltage exceeds 0.25V, the current is limited, and regeneration mode is established. In the application circuit, this voltage is set in such a way that the current limit will be established at 1A.

The calculation formula is given below.

RF resistance = 0.25V/target current limit value

*7. SOFTST pin

This pin is used to set the soft start.

By connecting a capacitor between this pin and GND, the motor speed can be increased gradually.

When the pin voltage exceeds 2.5V, the soft start is released, and the LV8805SV is switched to normal control.

If the soft start function is not going to be used, connect the pin to the VREG pin.

*8. OSC pin

This pin is used to connect the capacitor for setting the startup frequency.

A capacitor with a capacitance ranging from about 500pF to 2,200pF (recommendation) must be connected between this pin and GND.

The OSC pin determines the motor startup frequency, so be sure to connect a capacitor to it.

<How to select the capacitance>

Select a capacitance value that will result in the shortest possible startup time for achieving the target speed and produce minimal variations in the startup time. If the capacitance is too high, variations in the startup time will increase; conversely, if it is too low, the motor may idle. The optimum OSC constant depends on the motor characteristics and startup current, so be sure to recheck them when the type of motor used or circuit specifications are changed.

*9. VG, CP, and CPC pins

These pins are used to connect the capacitors to generate the pre-drive voltage and stabilize the pre-drive power supply.

Be sure to connect these capacitors in order to generate the drive voltage for the high-side (upper) output DMOS transistor.

*10. VREG pins

These are the control system power supply pin and regulator output pin, which create the power supply of the control unit. Be sure to connect a capacitor between this pin and GND in order to stabilize control system operation.

Since these pins are used to supply current for control and generate the charge pump voltage, connect a capacitor with a capacitance that is higher than that of the capacitor connected to the charge pump.

Both the VREG pins (pins 3 and 4) must be short-circuited on the print pattern.

*11. Pin protection resistor

It is recommended that resistors higher than $1k\Omega$ are connected serially to protect pins against misconnection such as GND open and reverse connection.

ORDERING INFORMATION

Device	Package	Shipping (Qty / Packing)	
LV8805SV-MPB-H	SSOP20J (225mil) (Pb-Free / Halogen Free)	90 / Fan-Fold	
LV8805SV-TLM-H	SSOP20J (225mil) (Pb-Free / Halogen Free)	2000 / Tape & Reel	

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